

New Cuts of Quartz for Surface Transverse Wave Temperature Wireless Sensors



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Abstract

A surface transverse wave (STW) in quartz is interesting for application at high frequencies because of higher velocity than that of a surface acoustic wave (SAW). STW propagates along the plane of Y rotated cut perpendicularly to X axis of quartz and is described as $\theta YX90^\circ$, where θ is rotation angle. The plane should be covered with periodical electrodes and period of electrodes p should be equal to $\lambda/2$, where λ is the STW wavelength. So far, the STW were investigated only for the temperature compensated cut at $\theta \cong 36^\circ$. In this work we experimentally investigated three cuts with expected high first order temperature coefficients of frequency (TCF_1). For $\theta \cong 43^\circ$ (ST cut for SAW propagating in the X direction), $TCF_1 \cong 35$ ppm/ $^\circ$ C, was obtained. However, the second order coefficient $TCF_2 \cong -53$ ppb/ $^\circ$ C² introduced relatively large nonlinearity to the frequency changes against temperature. For the new cuts at $\theta \cong 47^\circ$ and $\theta \cong 55^\circ$, the TCF_1 of about 45 ppm/ $^\circ$ C and 71 ppm/ $^\circ$ C respectively, were obtained. $TCF_2 = 0$ was obtained for both cuts. The results show that the STW on the new cuts of quartz are attractive for application in wireless temperature sensors.

Keywords: Quartz; Surface transverse wave; Wireless temperature sensor

Abbreviations: Surface Transverse Wave: STW; Surface Acoustic Wave: SAW

Introduction

The Surface Transverse Wave (STW) propagates along the plane of Y rotated cut perpendicularly to X axis of quartz, which is described as $\theta YX90^\circ$, where θ is the rotation angle. The plane should be covered with periodical electrodes and period of electrodes p should be equal to $\lambda/2$, where λ is the STW wavelength [1]. So far, the STW were investigated only for the temperature compensated cut ($\theta \cong 36^\circ$) [2]. For applications in the temperature sensors, new cuts, with sufficiently large linear temperature coefficients of frequency should be found. Because both Love and STW have only horizontal components of mechanical motion parallel to the X axis of quartz, it is expected that measured temperature coefficients of frequency will be similar to both waves [3]. Large velocity, lack of vertical component of mechanical motion (no air load losses compared to SAW [4]), larger penetration depth and possibility of large linear temperature coefficients of frequency, make the STW attractive for application in the temperature sensors at high frequencies.

Materials and Methods

Taking into account the calculated results for the first and second order temperature coefficients of frequency presented

in [3], the following three rotated Y-cut angles θ were chosen for measurements of the STW properties: 43° , 47° and 55° . The first cut is the well known ST cut used for SAW propagating in the X axis direction. The following general expressions for such STW parameters as phase velocity v , electromechanical coupling coefficient K^2 and the reflection coefficient γ of a single aluminum electrode, can be written as [5]:

$$v = v_o [1 - 14.8(h/\lambda)^2] \quad (1)$$

$$K^2 = a(h/\lambda) \quad (2)$$

$$\gamma = b(h/\lambda)^2 \quad (3)$$

where h is the aluminum layer thickness and λ is the STW wavelength. Velocity v_o and the coefficients a and b depend on the cut angle of quartz.

Photolithographic mask (Figure 1), designed for the STW resonator on the temperature compensated cut of quartz ($\theta = 36^\circ$) [5] was used in the present work. Here $W = 1$ mm, $p_1 = 5$ μ m, $p_2 = 4.9$ μ m, $p_3 = 4.98$ μ m, $\lambda = 2p_1 = 10$ μ m. The numbers of electrodes for the reflector, for the phase shifter, for the IDT and for the centre grating were equal to 800, 45, 121 and 225, respectively.

The resonators were fabricated by the lift-off method for the aluminum layer thickness $h = 0.1 \mu\text{m}$ ($h/\lambda = 1\%$) and measured (Agilent Technologies Network Analyzer Type 8753ET). As an example, amplitude response of a resonator on the $\theta = 55^\circ$ cut quartz is shown in Figure 2. Using the algorithm presented in

[2], the STW parameters were changed until the measured and calculated amplitude responses were matched. The obtained parameters are shown in Table 1. These parameters can be used for design of the STW resonators.

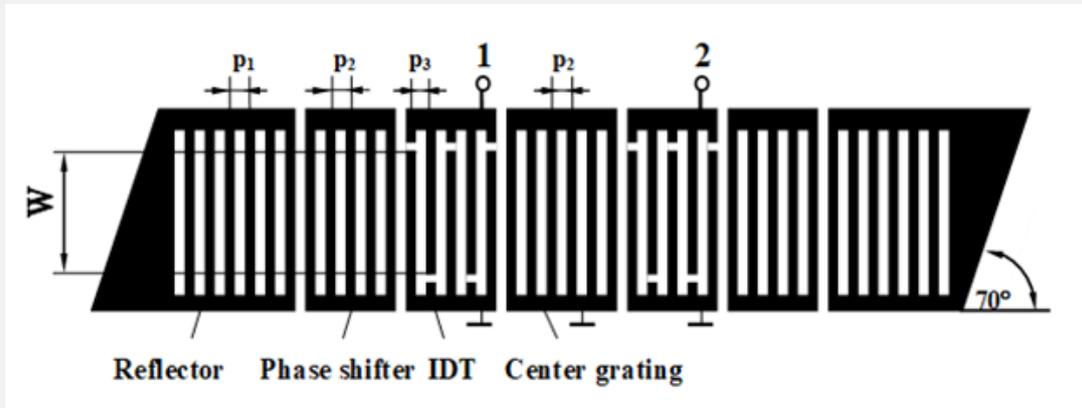


Figure 1: Structure of asynchronous STW resonator.

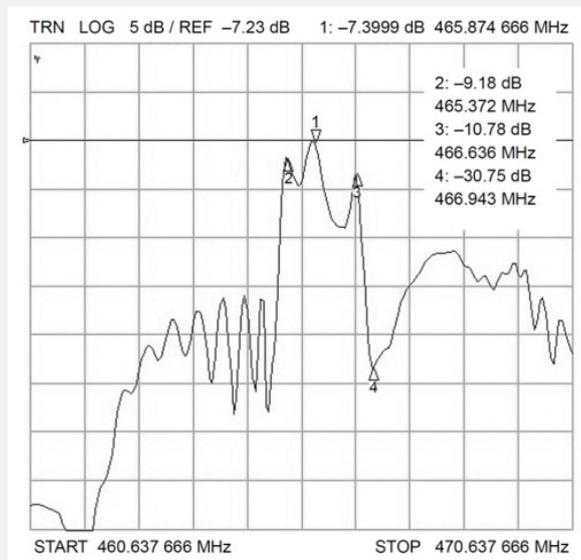


Figure 2: Measured amplitude response of STW resonator on 55o cut.

Results

The STW resonators were located in the temperature test chamber and the relative changes of frequency against temperature were measured. The result of measurements are shown in Figure 3. A temperature dependence of frequency $f(T)$ can be written as:

$$f(T) = f_0 [1 + TCF_1(T - T_0) + TCF_2(T - T_0)^2] \quad (4)$$

where T_0 is the reference temperature and f_0 is the frequency at $T = T_0$.

The first and second order temperature coefficients of frequency defined as:

$$TCF_1 = \Delta f / (f_0 \Delta T) \quad (5)$$

$$TCF_2 = \Delta f / (f_0 \Delta T^2) \quad (6)$$

were determined from the measurements and are presented in Table 2.

It can be seen that for the 43° cut significant TCF_2 exists, while for the new cuts this coefficient is equal to zero. It is expected that the property will also exist between the 47° and 55° cut angles.

To verify the above results, a resonator for the first ISM band at a frequency of about 434 MHz, was designed and fabricated on the 55° cut. Using the STW parameters determined previously (Table 1) and the design guidelines presented in [5], the following data of the resonator structure were obtained: $W = 1\text{ mm}$, $p_1 = 5.37$

μm , $p_2 = 5.31\ \mu\text{m}$, $p_3 = 5.35\ \mu\text{m}$ and $\lambda = 2p_1 = 10.74\ \mu\text{m}$. The numbers of electrodes for the reflector, for the phase shifter, for the IDT and for the centre grating were equal to 600, 177, 141 and 220, respectively. The obtained amplitude response of the resonator, for the aluminum layer $h = 0.1\ \mu\text{m}$, is shown in Figure 4.

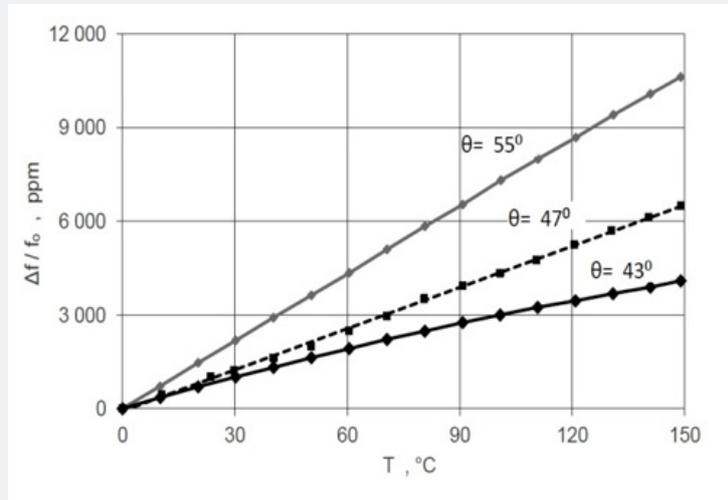


Figure 3: Measured relative changes of frequency against temperature.

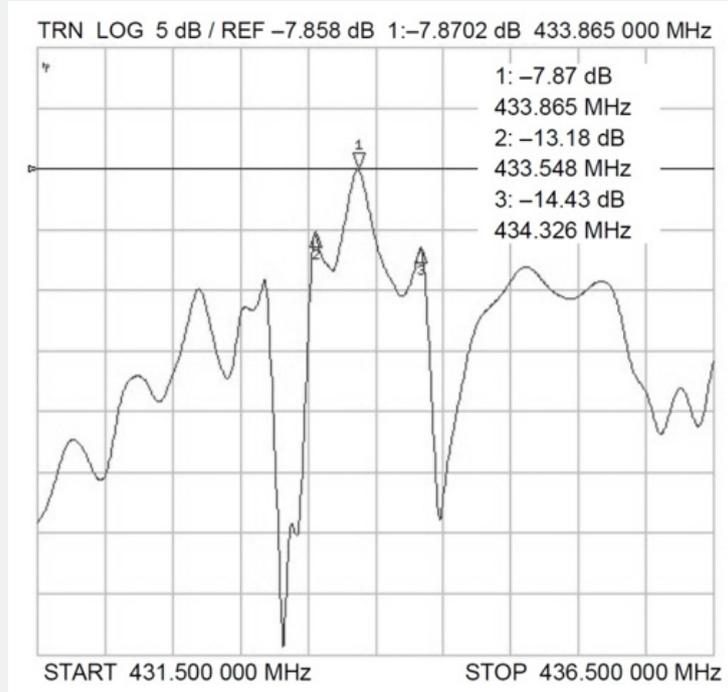


Figure 4: Measured amplitude response of STW resonator on 55o cut for the first ISM band.

Taking into account the internal insertion loss of the measuring system of about 2 dB, the net insertion loss 6 dB and the center frequency 433.865 MHz, were obtained. The loss coefficient $T_i \cong 0.754$ was used to match the calculated and measured net insertion

loss and loaded QL and unloaded QU quality factors of about 3400 and 7000 respectively, were obtained to verify the above results, a resonator for the first ISM band at the frequency of about 434 MHz, was designed and fabricated on the 55° cut. Using the STW

parameters determined previously (Table 1) and the design guidelines presented in [5], the following data of the resonator structure were obtained: $W = 1\text{mm}$, $p_1 = 5.37\ \mu\text{m}$, $p_2 = 5.31\ \mu\text{m}$, $p_3 = 5.35\ \mu\text{m}$ and $\lambda = 2p_1 = 10.74\ \mu\text{m}$. The numbers of electrodes for the reflector, for the phase shifter, for the IDT and for the centre grating were equal to 600, 177, 141 and 220, respectively. The obtained amplitude response of the resonator, for the aluminum layer $h = 0.1\ \mu\text{m}$, is shown in Figure 4.

Table 1: Parameters of STW for different cuts.

Cut Angle θ	43°	47°	55°
v_0 [m/s]	4999	4884	4633
a	0.27	0.34	0.34
b	-43	-52	-52

Table 2: Measured temperature coefficients of frequency.

Cut Angle θ	43°	47°	55°
TCF_1 [ppm/°C]	35	45	71
TCF_2 [ppb/°C ²]	-53	0	0

Taking into account the internal insertion loss of the measuring system of about 2 dB, the net insertion loss 6 dB and the center frequency 433.865 MHz, were obtained. The loss coefficient $T_i \cong 0.754$ was used to match the calculated and measured net insertion loss and loaded $Q_L \cong 3400$ and unloaded $Q_U \cong 7000$ quality factors.

Discussion

Both structure and design of the STW asynchronous resonator is complicated because of different period of electrodes in the IDTs, in the center grating, and in the phase shifters [5]. For lower quality factors, simpler STW components can be used. For example, amplitude response of a STW delay line composed of 3 IDTs, was first presented in [3]. Internal reflections inside the IDTs and reflections between them changed the shape of the amplitude response and low insertion loss was obtained. Another possibility is a STW filter [6]. The filter consists of two long IDTs and a short

grating shield located between them. Period of electrodes is equal to $\lambda/2$ and is constant along the whole length of both components. So far, one-port STW resonator was not investigated.

STW can also be used in the wireless passive sensors [7]. In this case, the distances between reflectors should be covered by metal gratings with period of electrodes smaller than $\lambda/2$ [5].

Conclusion

Such STW parameters as velocity, electromechanical coupling coefficient, reflection coefficient of a single aluminum electrode and temperature coefficients of frequency were determined experimentally for the 43°, 47° and 55° cut angles of Y rotated quartz. High velocity, high sensitivity and good linearity of the relative frequency changes against temperature was obtained for the last two cuts. The obtained results show that the STWs on the new cuts of quartz are attractive for application in wireless temperature sensors at high frequencies.

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